

Variable Threshold Voltage CMOS in Series Connected Circuits

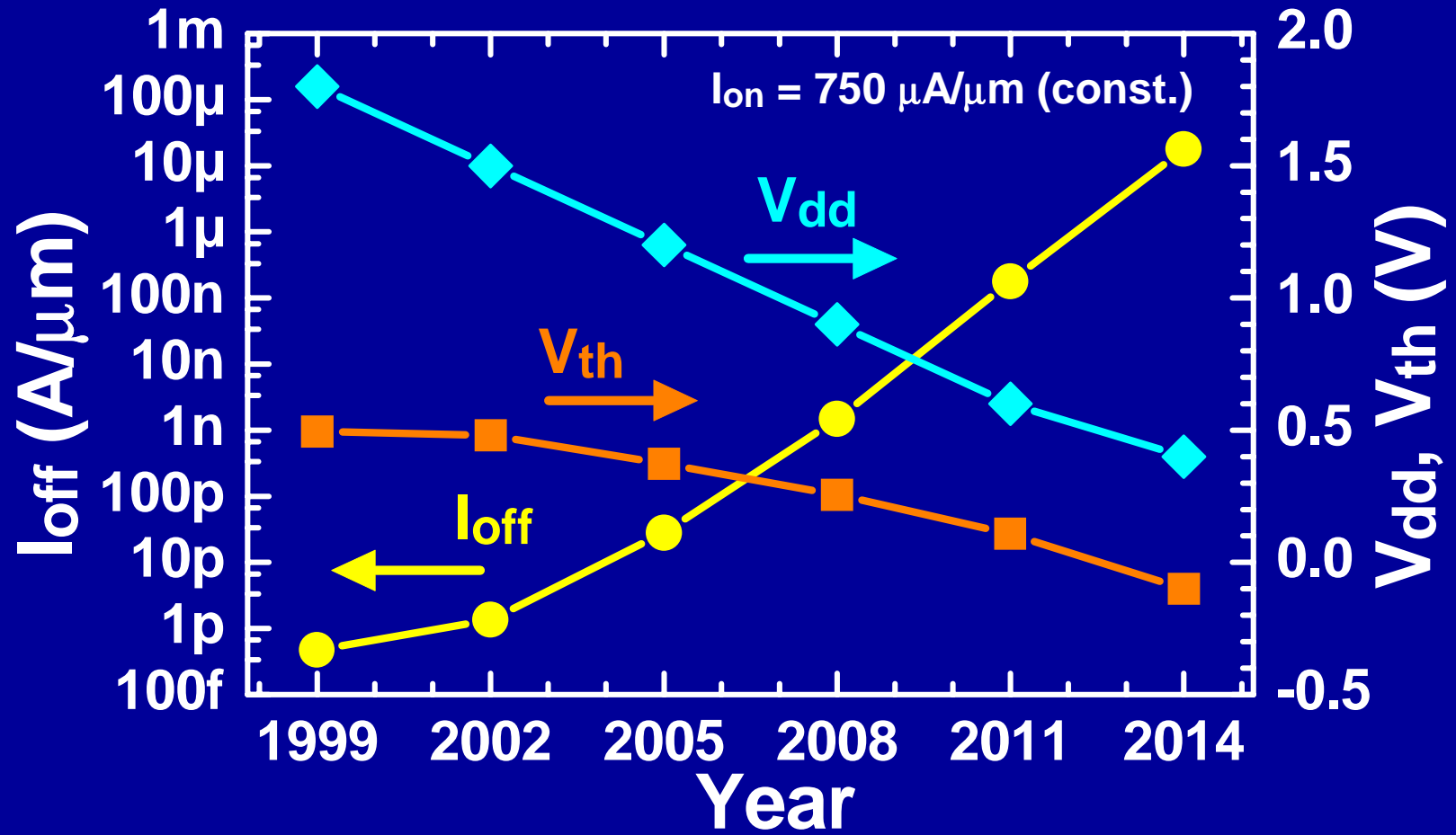
T. Inukai, T. Hiramoto and T. Sakurai

***Institute of Industrial Science,
the University of Tokyo, Japan***

Outline

- **Background and objective**
- **Body effect factor and its relationship with device characteristics**
- **Device design in VTCMOS**
 - **Single devices**
 - **Series connected circuits**
- **Effect of velocity saturation**
- **Summary**

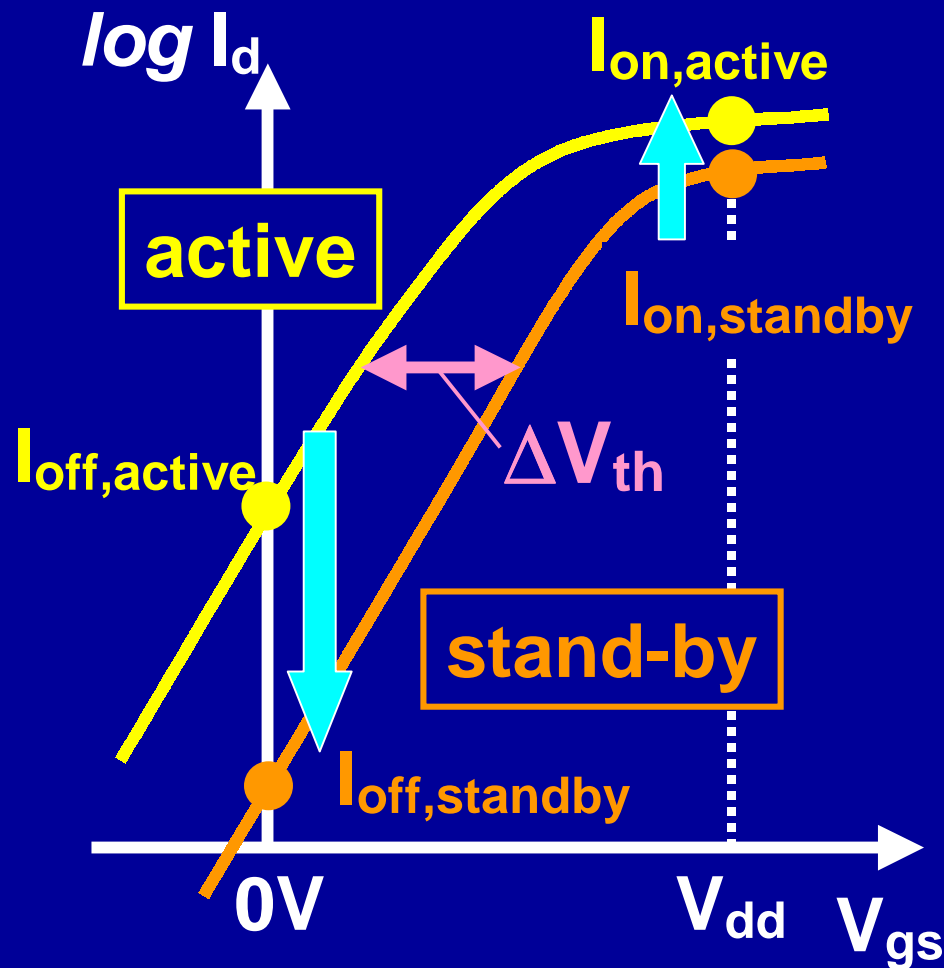
Explosion of Stand-by Power



Scaling of V_{th} with V_{dd} causes an increase of off-current.

➔ **Circuits with single (or fixed) V_{th} will fail due to explosion of stand-by power consumption.**

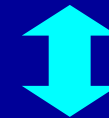
Variable Threshold Voltage CMOS



Utilizing the body effect, V_{th} is controlled by V_{bs} .

Low V_{th} @ Active mode

→ enhancement of I_{on}



High V_{th} @ Standby mode

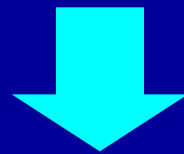
→ suppression of I_{off}

What determines ΔV_{th} ?

[ref.] T. Kuroda *et al.* IEEE JSSC, 31, 1770 (1996).

Objective

- Investigation of optimum device design in variable threshold voltage CMOS (VTCMOS) by means of device simulation from the viewpoints of
 - Device performance (**single device**)
 - Circuit performance (**series connected circuit**)



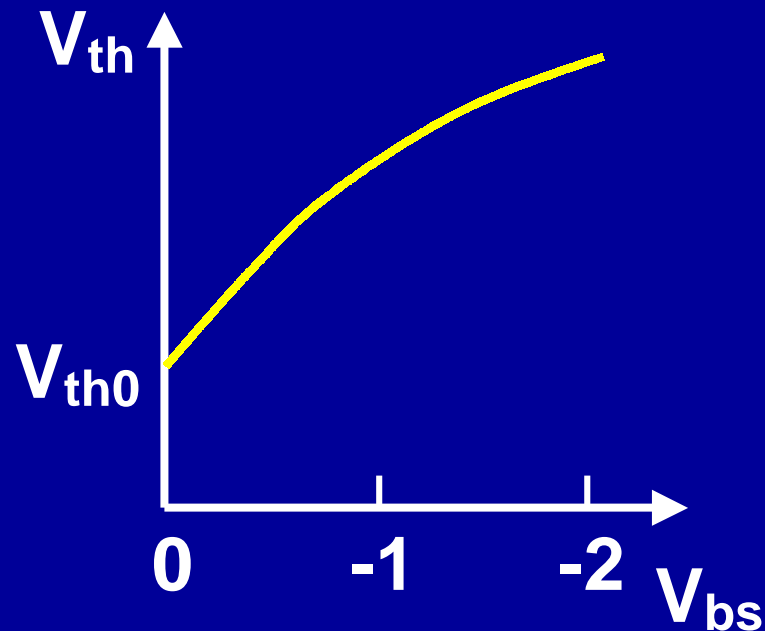
Guideline for device design in VTCMOS

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Body Effect Factor

V_{th} is varied by the body effect



$$V_{th} = 2\phi_F + \frac{\sqrt{2q\epsilon N_a(2\phi_F + |V_{bs}|)}}{C_{ox}}$$

$$\Delta V_{th} = \frac{\sqrt{2q\epsilon N_a}}{C_{ox}} (\sqrt{2\phi_F + |V_{bs}|} - \sqrt{2\phi_F})$$

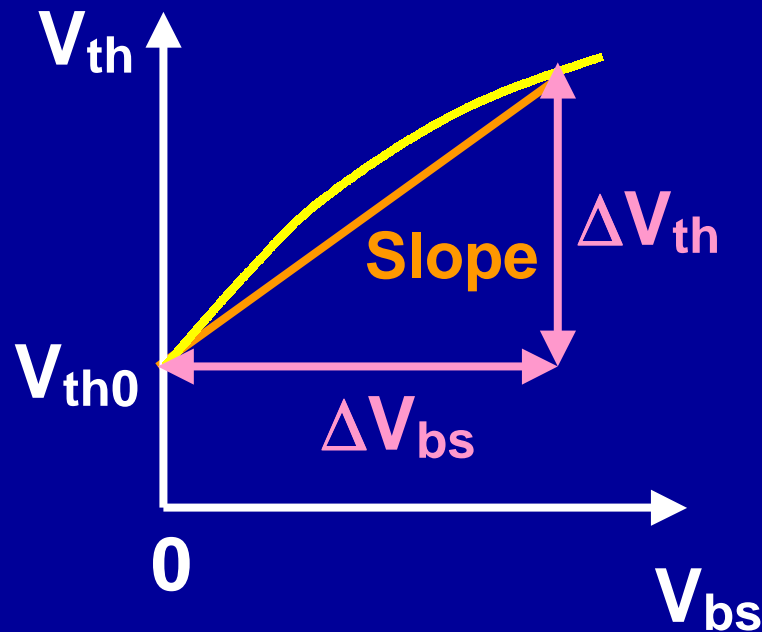


$$\gamma_{conv} \equiv \frac{\sqrt{2q\epsilon N_a}}{C_{ox}}$$

We did not use this definition!

- Valid only for **uniformly-doped MOSFET**
- Hard to estimate ΔV_{th} (Unit is $V^{1/2}$)

Definition of Body Effect Factor



$$\gamma \equiv \frac{|\Delta V_{th}|}{|\Delta V_{bs}|}$$

Then, $\Delta V_{th} = \gamma |\Delta V_{bs}|$

Moreover, $\gamma \approx \frac{C_d}{C_{ox}}$

(C_d : depletion cap., C_{ox} : oxide cap.)

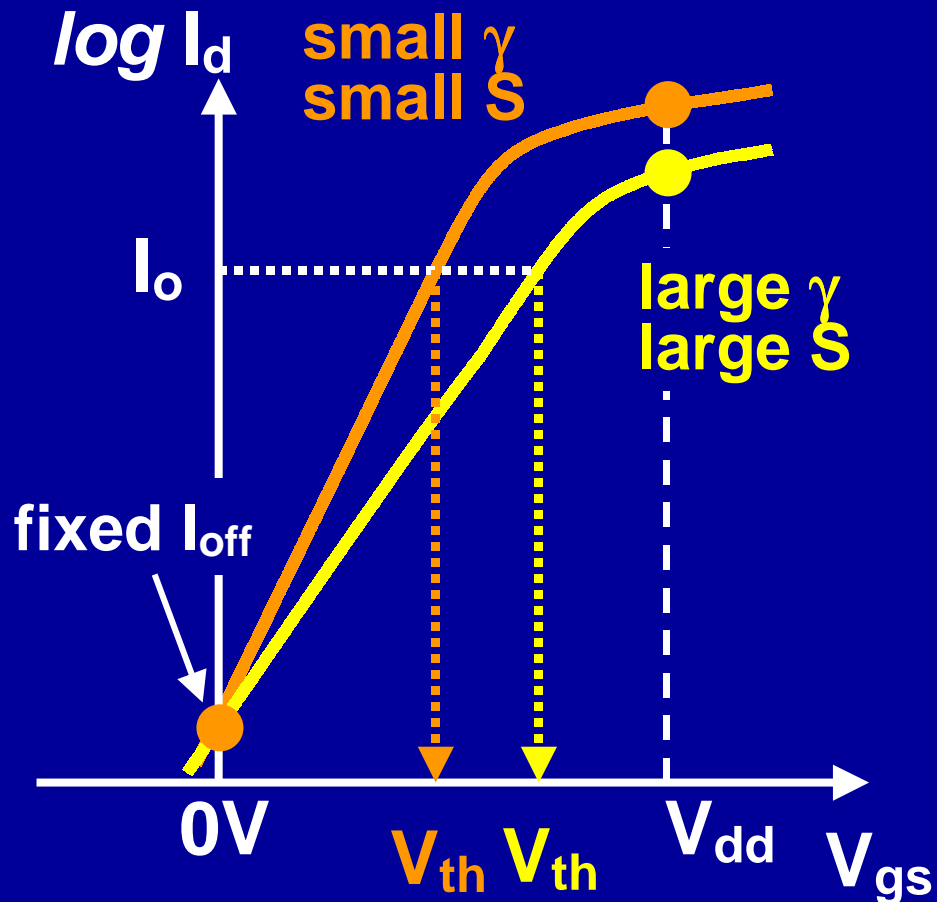
Good approximation in

- super steep retrograde profile
- very low ΔV_{bs}

- Directly related to ΔV_{th}
- Applicable to any devices
- Also related to device performance

Relation between Device Performance and γ (1)

Subthreshold characteristic



(for long channel device)

$$\begin{cases} S = 60 \left(1 + \frac{C_d}{C_{ox}} \right) \\ \gamma \approx \frac{C_d}{C_{ox}} \end{cases}$$

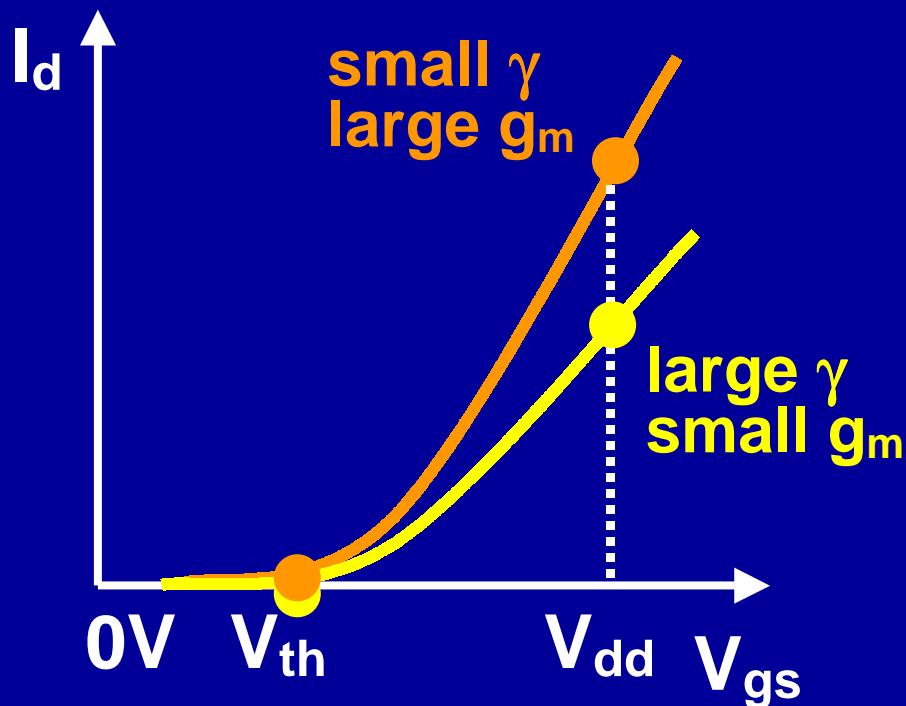
→ **$S \approx 60(1 + \gamma)$**

	small γ	large γ
S factor	small	large
V_{th}^*	low	high

* for fixed I_{off}

Relation between Device Performance and γ (2)

Transconductance



(for long channel device)

$$I_{dsat} = \mu_{eff} C_{ox} \frac{W}{L} \frac{(V_{gs} - V_{th})^2}{2(1+\gamma)}$$



Charge loss due to the body effect

	small γ	large γ
g_m	large	small
I_{on}^*	high	low

* for fixed V_{th}

Trade-off between V_{th} Control and Device Performance

Larger γ device can achieve larger ΔV_{th}



trade-off

Larger γ device has lower I_{on} due to

- larger S factor (\rightarrow higher V_{th})
- smaller transconductance (g_m)



How should we design the device in VTCMOS?

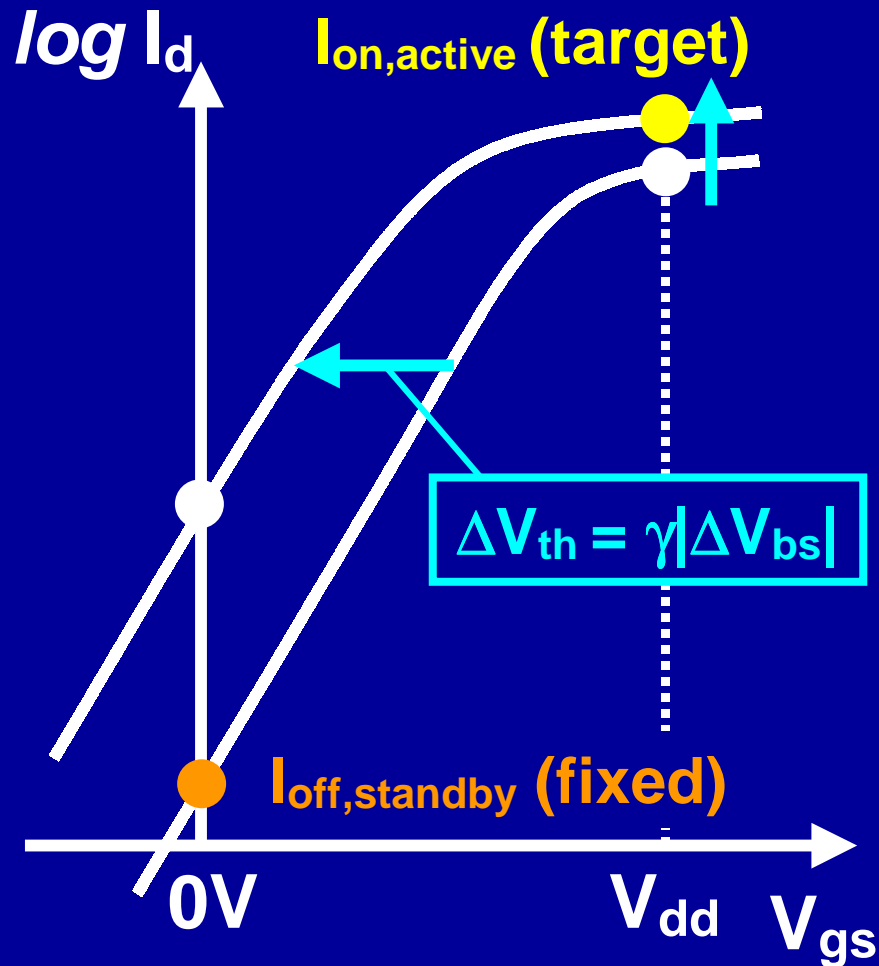


Systematical investigation by means of device simulation

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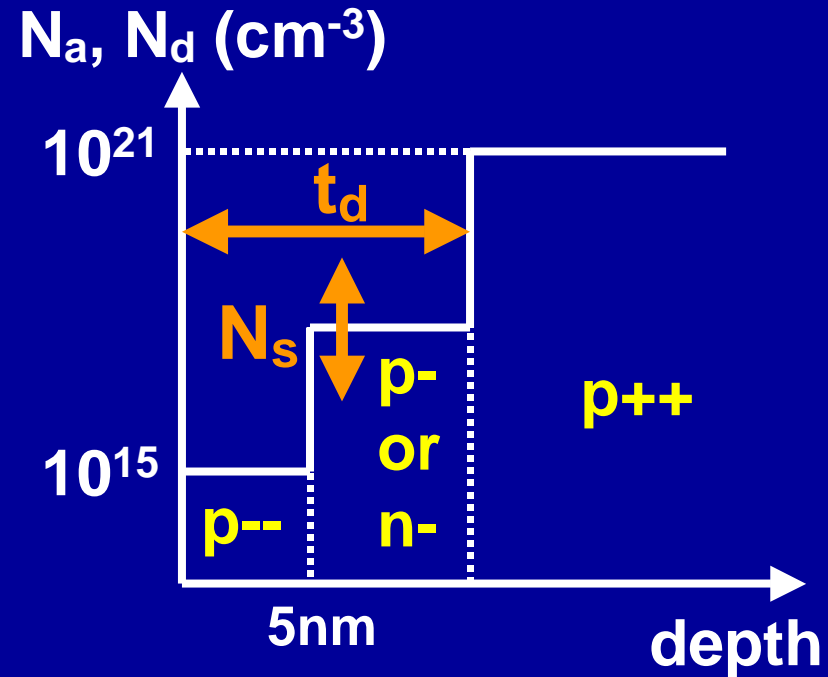
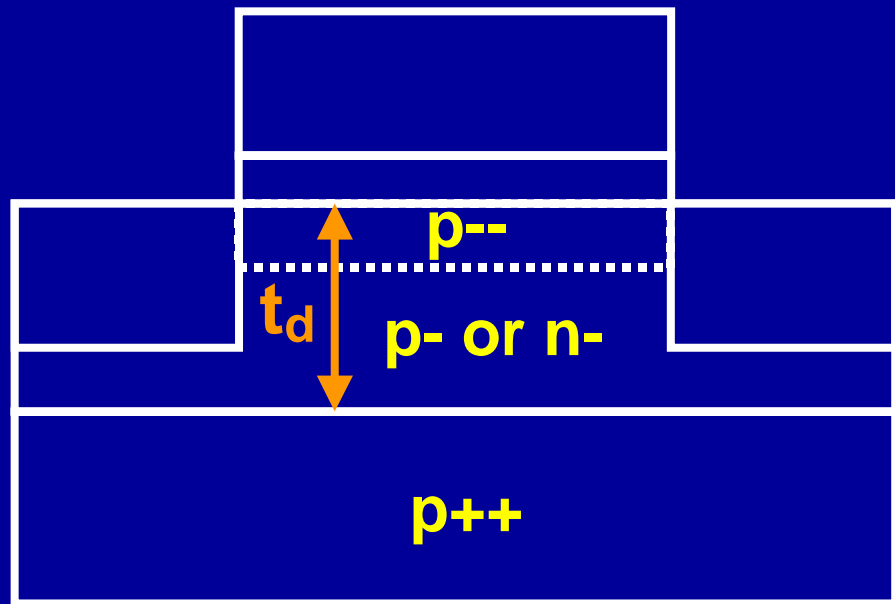
Simulation Procedure



- With fixed $I_{off,standby}$, $I_{on,active}$ is compared.
- γ and $|\Delta V_{bs}|$ are varied separately.

How can we achieve higher on-current?

Device Structure



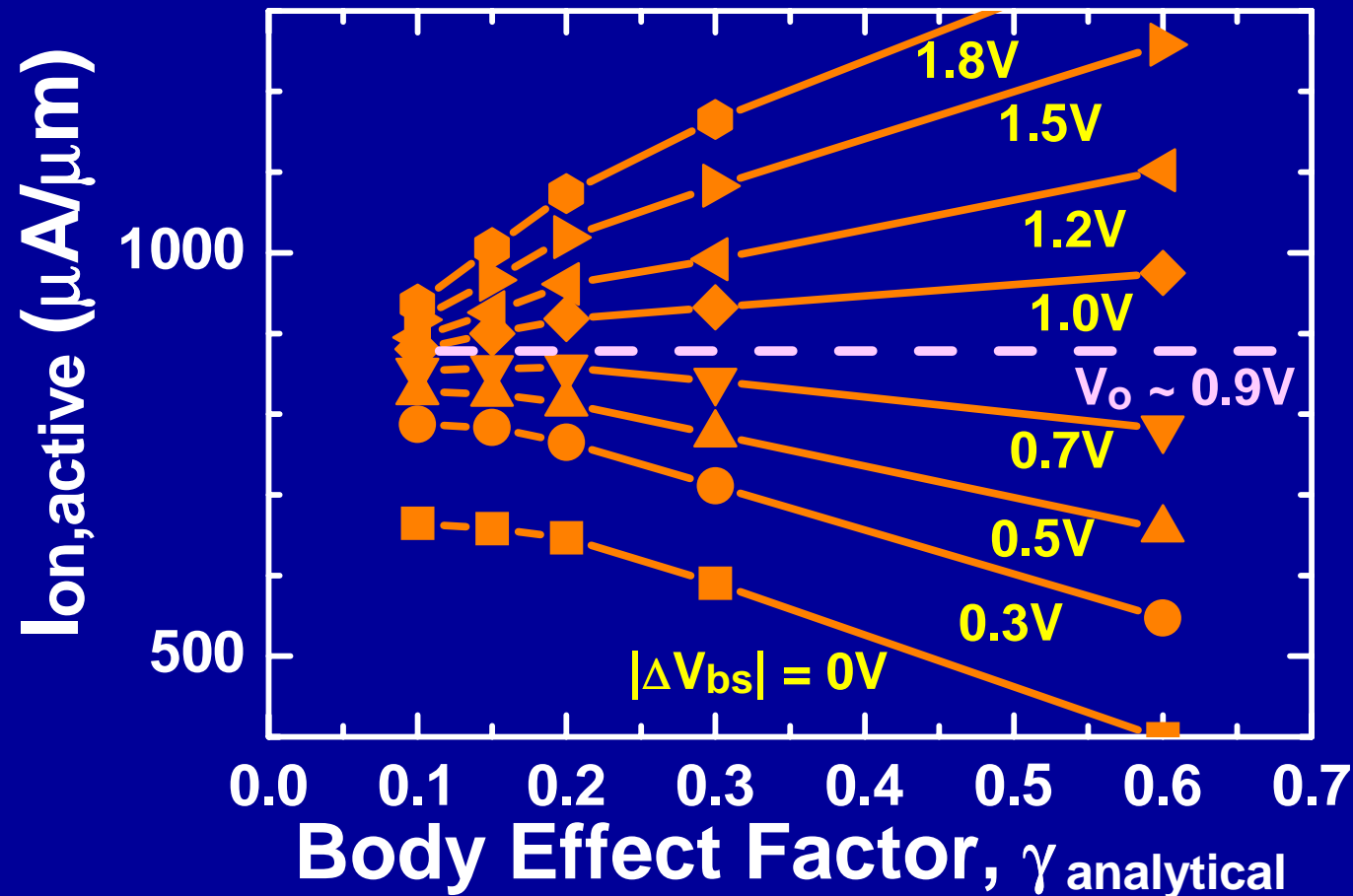
Ideal step-shape channel profile

- t_d is varied to change γ .
- N_s is varied to change V_{th} so that $I_{off,standby}$ would be fixed for given $|\Delta V_{bs}|$.
- Non-doped layer (5nm thick) is inserted to avoid inaccuracy of model for impurity dependence of mobility.

$$\gamma_{\text{analytical}} = \frac{C_d}{C_{ox}} \approx 3 \frac{t_{ox}}{t_d}$$

γ Dependence of *On*-current

$I_{\text{off, standby}} = 0.1 \text{ pA}/\mu\text{m}$ (const.)



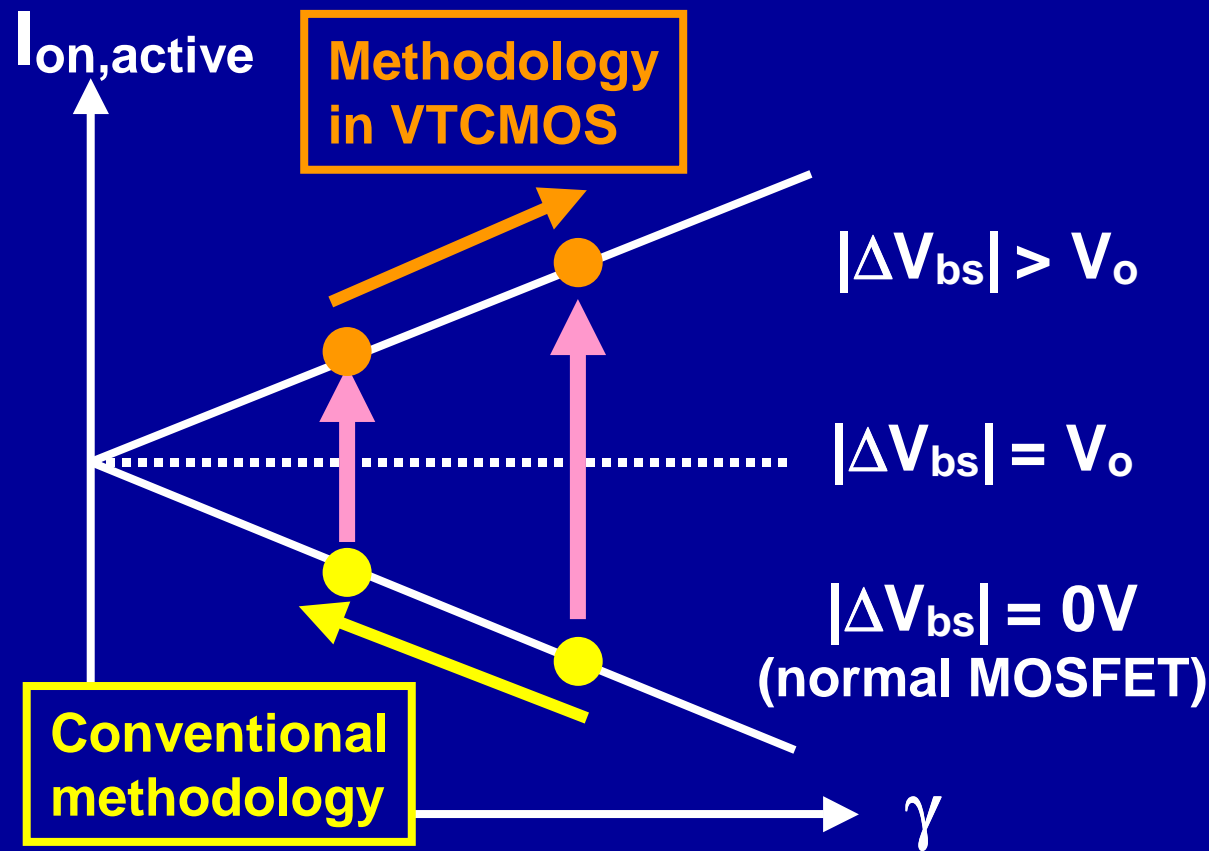
When $|\Delta V_{\text{bs}}| < V_o$, larger γ device has lower $I_{\text{on, active}}$.



When $|\Delta V_{\text{bs}}| > V_o$, larger γ device has higher $I_{\text{on, active}}$.

Design Methodology in VTCMOS

- Single Devices -

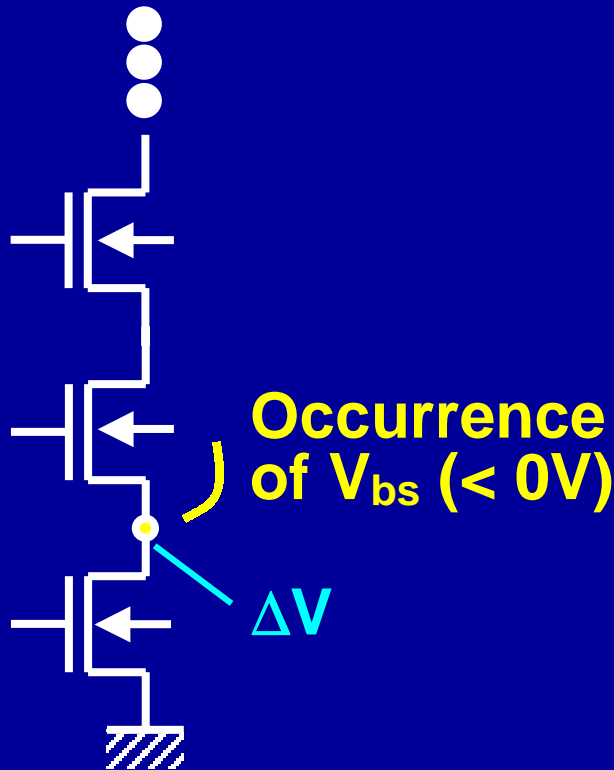


Contradictory to the conventional methodology, γ should be as large as possible in VTCMOS when $|\Delta V_{bs}| > V_o$ from the viewpoints of $I_{on,active}$.

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Series Connected Circuits



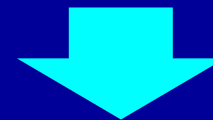
Some source nodes are not connected to GND lines but floating.



Negative substrate bias ($V_{bs} < 0V$) occurs even in **active mode**.



Circuit performance is deteriorated by **the body effect**.



Advantage of large γ device in VT CMOS might be cancelled out in series connected circuits.



Further investigation in consideration of the series connected circuits

Evaluation of Performance in Series Connected Circuits

Degradation Factor : F_d

$$F_d \equiv \frac{I_{on}}{I_{on,N}}$$

I_{on} : *on*-current of single device

$I_{on,N}$: *on*-current of series connected circuit

$$= 1 + \frac{1 - 1/\sqrt{2}}{1 - 1/\alpha\sqrt{2}} \frac{V_{dsat}}{V_{dd} - V_{th}} (1 + \gamma)(1 + \lambda V_{dd})(N - 1)$$

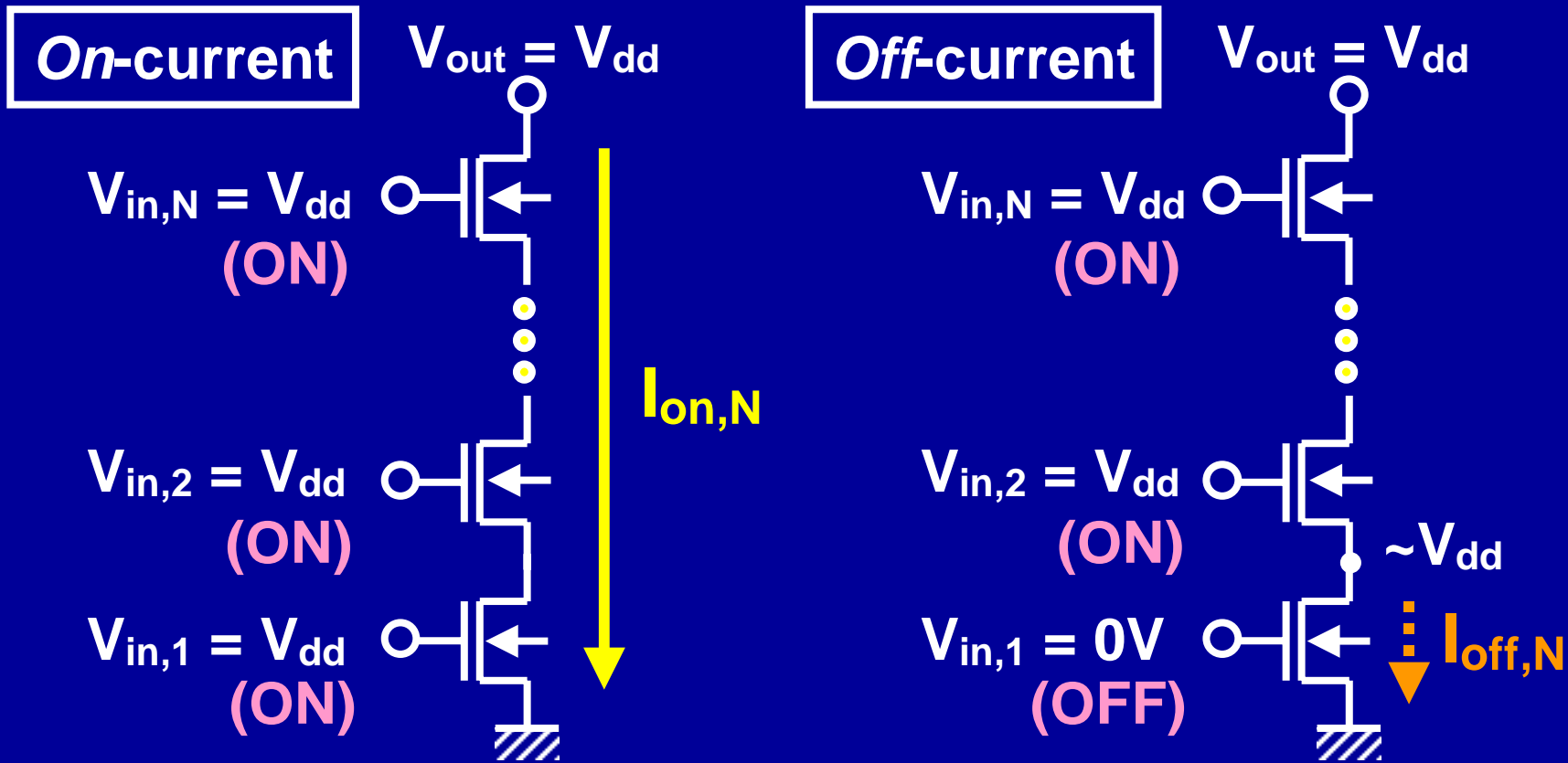
$$\approx 1 + \frac{1}{2} \alpha \frac{V_{dsat}}{V_{dd} - V_{th}} (1 + \gamma)(1 + \lambda V_{dd})(N - 1)$$



Degradation due to the body effect

[ref.] T. Sakurai *et al.* IEEE JSSC, 26, 122 (1991).

On- and Off-currents in Series Connected Circuits

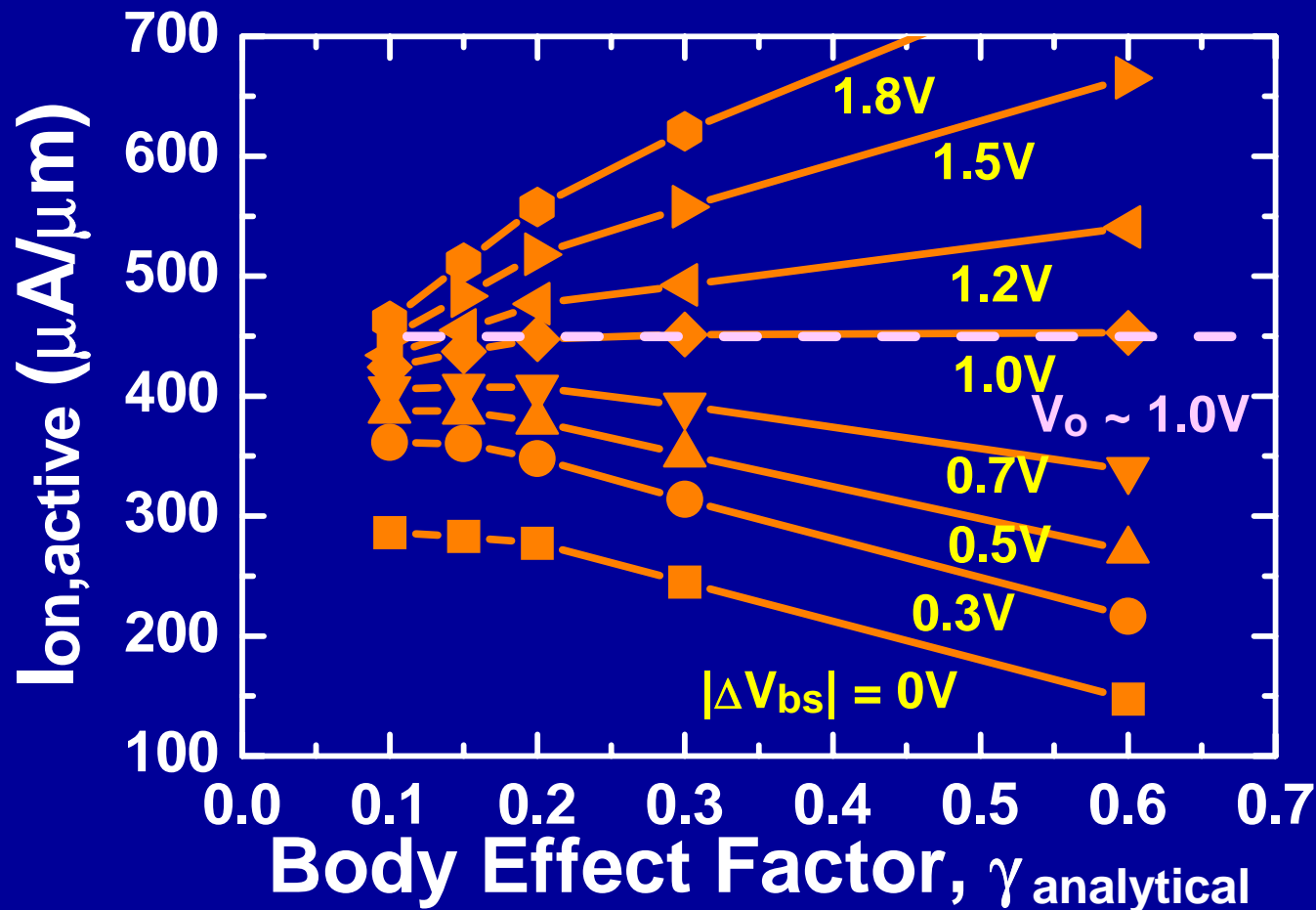


Worst case *off*-current is almost same as *off*-current of single devices.

➔ Same boundary condition (fixed $I_{off,stanby}$) can be used.

γ Dependence of I_{on} -current in Series Connected Circuits

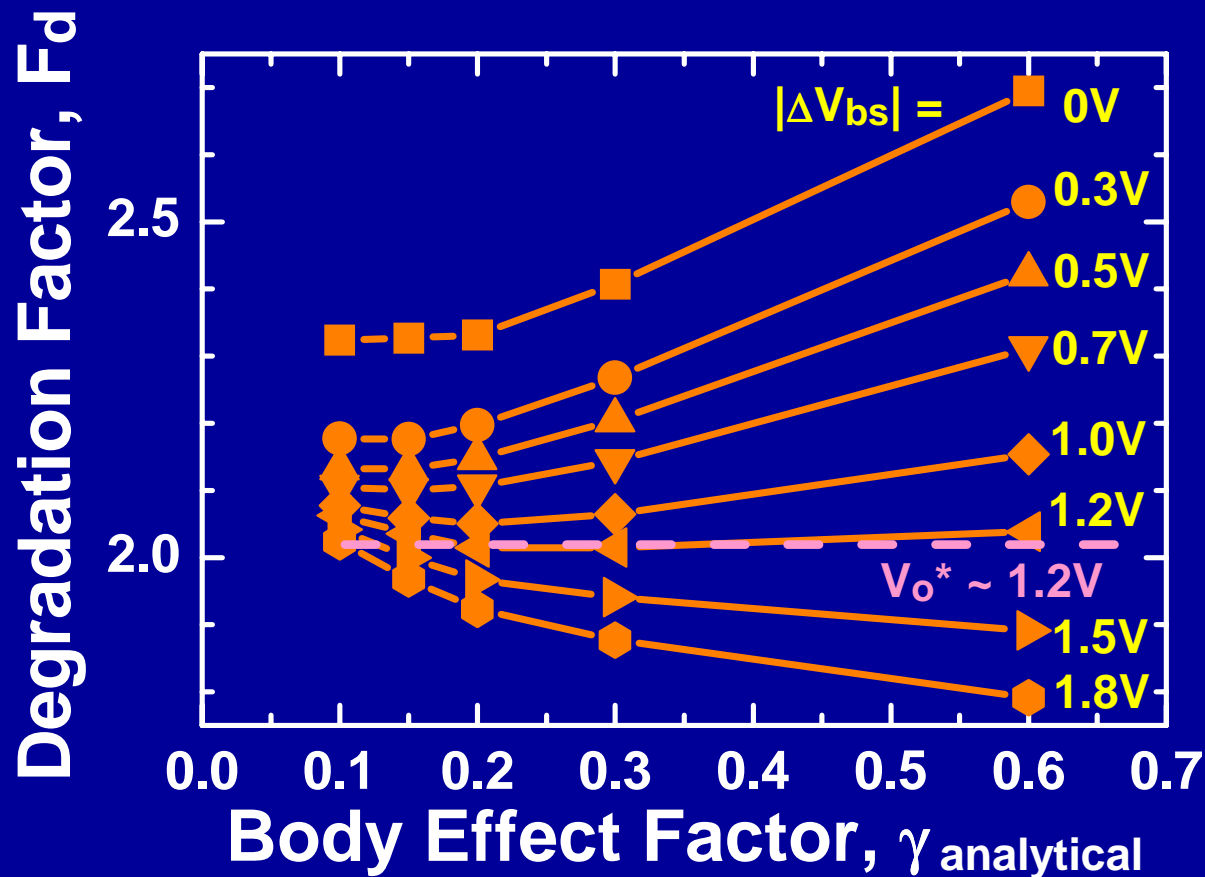
$I_{off,standby} = 0.1 \text{ pA}/\mu\text{m}$ (const.)



of Tr. = 3

Although V_o becomes higher (from 0.9V to 1.0V),
larger γ device keeps its advantage when $|\Delta V_{bs}| > V_o$.

γ Dependence of F_d



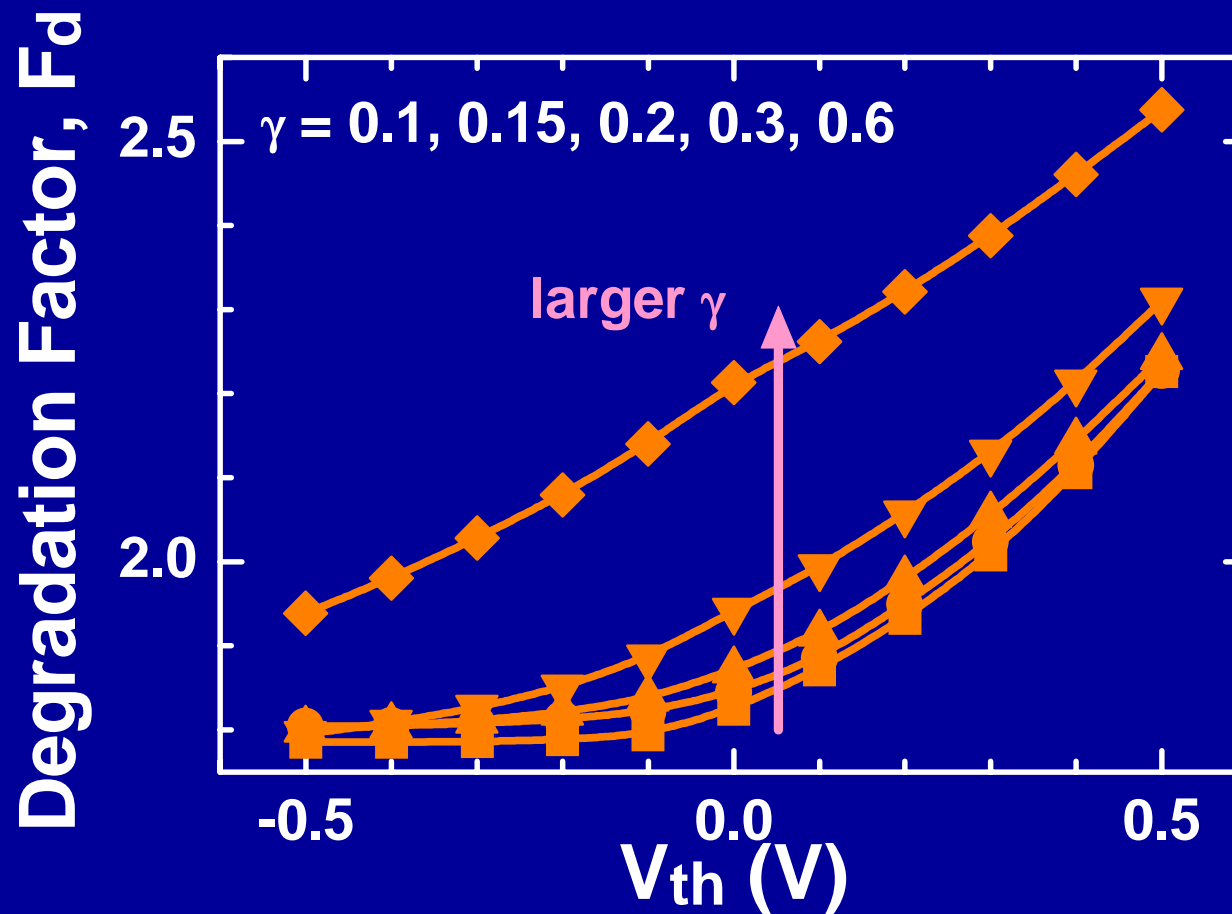
of Tr. = 3

$$F_d \equiv \frac{I_{on}}{I_{on,N}}$$

- Utilizing VTCMOS, F_d decreases in each γ value.
- As $|\Delta V_{bs}|$ increases above another critical value ($V_o^* \sim 1.2V$), γ dependence of F_d changes.

➔ Larger γ device has smaller F_d !?

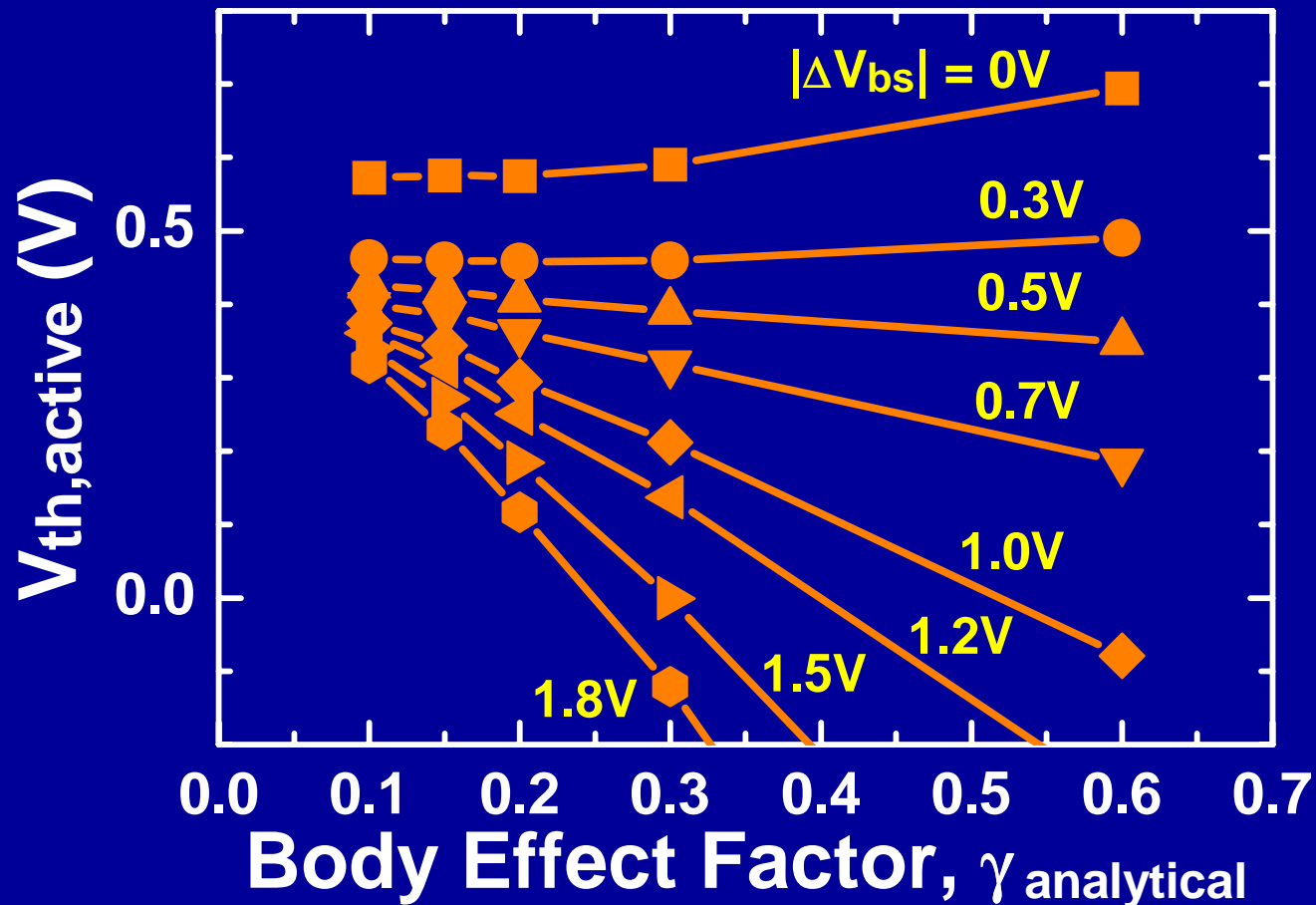
Effect of Lowering V_{th}



- With the constant V_{th} , larger γ has larger F_d .
- In any γ values, lowering V_{th} results in decrease of F_d .

γ Dependence of Active V_{th}

$I_{off,standby} = 0.1 \text{ pA}/\mu\text{m}$ (const.)



Larger γ device can achieve lower V_{th} in active mode with the fixed stand-by off-current.

➔ *Smaller F_d in spite of larger body effect*

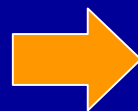
Design Methodology in VTCMOS

- Series Connected Circuits -

On-current in the series connected circuits

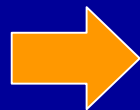
$$I_{on,N} = \frac{I_{on}}{F_d}$$

VTCMOS can not only **enhance** I_{on} but also **reduce** F_d by lowering V_{th} .



Larger enhancement of I_{on} in the series connected circuits

Larger γ device can achieve lower $V_{th,active}$, resulting in smaller F_d .

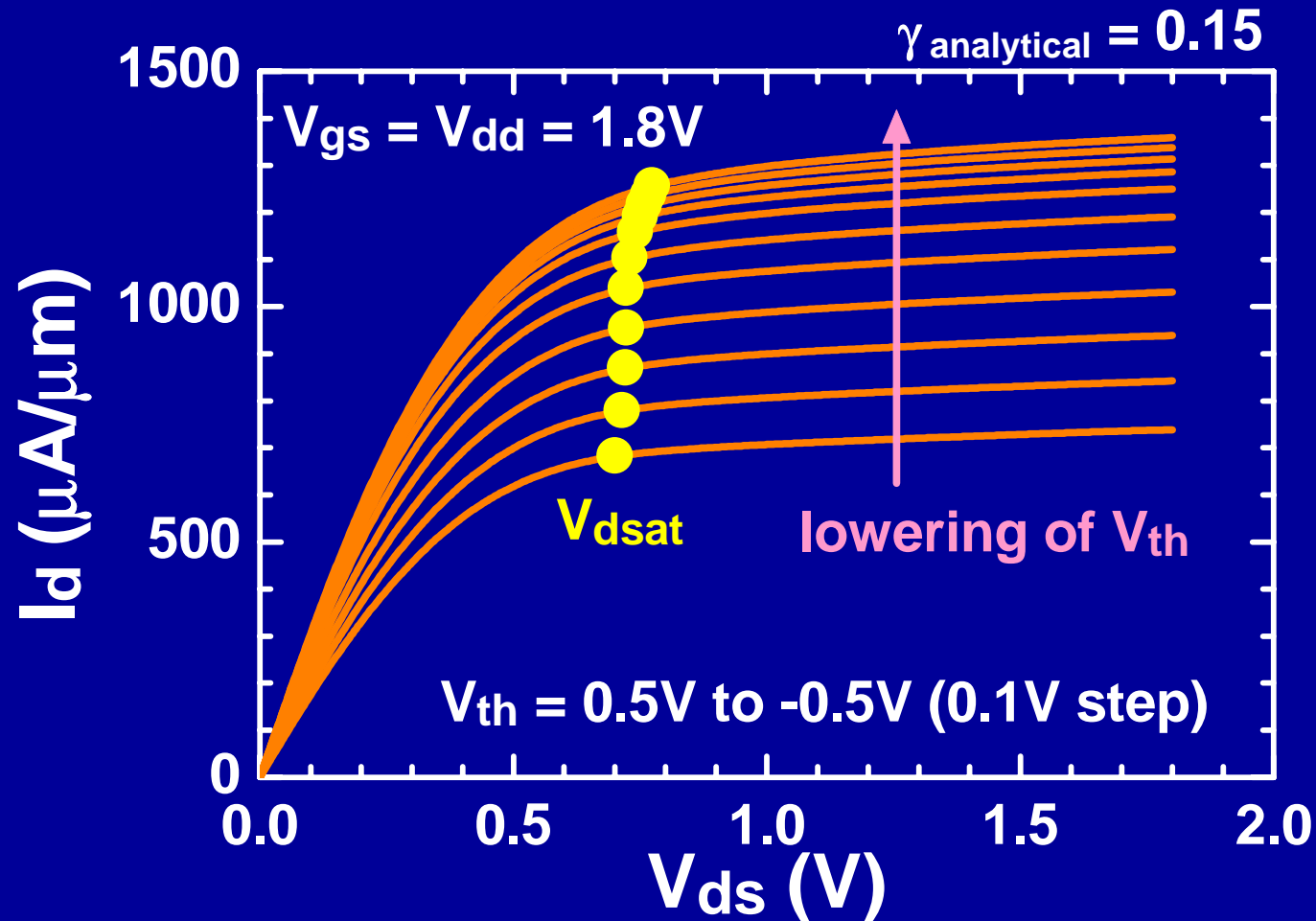


Advantage of large γ device in VTCMOS is not cancelled out but rather enhanced in the series connected circuits.

Outline

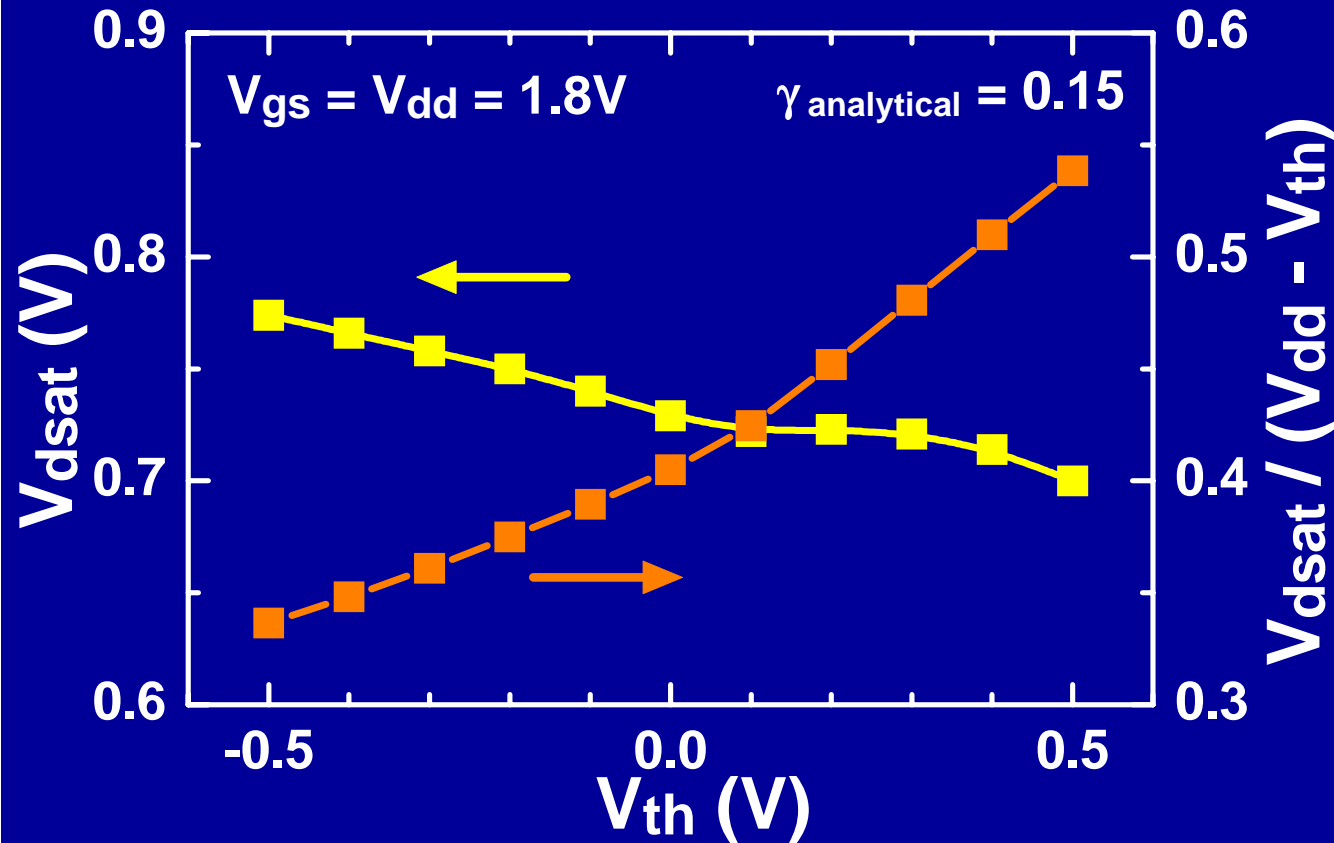
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Effect of Velocity Saturation



As V_{th} decreases, V_{dsat} **only slightly increases** due to **velocity saturation phenomenon**.

Mechanism of F_d Reduction



Slight increase
of V_{dsat} with
lowering V_{th}



$V_{dsat} / (V_{dd} - V_{th})$
decreases



F_d decreases !!

$$F_d \approx 1 + \frac{1}{2} \alpha \frac{V_{dsat}}{V_{dd} - V_{th}} (1 + \gamma)(1 + \lambda V_{dd})(N - 1)$$

Summary

- VTCMOS characteristic is systematically investigated by means of device simulation.
- From the viewpoint of device performance, larger γ device can achieve higher *on*-current, when $|\Delta V_{bs}|$ is larger than a certain critical value.
- Lowering V_{th} alleviates the degradation due to the series connected configuration. Therefore, when $|\Delta V_{bs}|$ is applied to some extent, larger γ device has higher *on*-current even in the series connected circuits.
- These characteristics are attributed to the velocity saturation phenomenon which makes the drain saturation voltage less sensitive to V_{th} .